

APPLICATION UNDER UNITED STATES PATENT LAWS

Invention: **SYNCHRONIZING DATA TRANSFER PROTOCOL ACROSS
HIGH VOLTAGE INTERFACE**

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This is a:

- ☐ Provisional Application
- ☒ Regular Utility Application
- ☐ Continuing Application
- ☐ PCT National Phase Application
- ☐ Design Application
- ☐ Reissue Application
- ☐ Plant Application

SPECIFICATION

SYNCHRONIZING DATA TRANSFER PROTOCOL ACROSS HIGH VOLTAGE INTERFACE

BACKGROUND OF THE INVENTION

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1. Field of the Invention

This invention relates to an improved serial interface. More particularly, it relates to the elimination of a synchronization signal such as a frame sync signal in a synchronized serial data stream or a time division
10 multiplex serial data stream by replacement with a synchronizing data protocol.

2. Background of Related Art

Serial communications are an effective method of
15 communicating between two digital components, particularly in cost sensitive applications to minimize the hardware requirements otherwise required for parallel communications. For instance, a serial interface technique implementing a transmit line, receive line, data clock line, and a frame synchronizing line (and a reset line) has been implemented in
20 conventional multipurpose codecs.

A codec is a device which for many years has allowed efficient and inexpensive digitization of telephone grade audio. The typical codec (short for COder-DECoder) is an integrated circuit or other electronic device which combines the circuits needed to convert analog
25 signals to and from digital signals, e.g., Pulse Code Modulation (PCM) digital signals.

Early codecs converted analog signals at an 8 KHz rate into 8-bit PCM for use in telephony, and were not capable of handling modem inputs. More recently, the efficiency and low cost advantages of codecs
30 have been expanded to convert analog signals at a 48 KHz sampling rate into 16-bit stereo (and even up to 20-bit stereo) for higher quality use

beyond that required for telephony. With higher quality and broader bandwidth capability, today's codecs find practical application with consumer equipment such as voice band modems.

With the development of codecs for these more
5 sophisticated purposes came the need to improve the analog signal-to-noise (S/N) ratio to at least 75 to 90 dB. One major step toward achieving this high S/N ratio was accomplished more recently by separating the conventional codec into two individual sub-systems: a controller sub-system or integrated circuit (IC) handling primarily the digital interface to a
10 host processor, and an analog sub-system or IC handling primarily the interface to, mixing and conversion of analog signals. This split digital/analog architecture has been documented most recently as the "Audio Codec '97 Component Specification", Revision 1.03, September 15, 1996 ("the AC '97 Specification"). The AC '97 Specification in its
15 entirety is expressly incorporated herein by reference.

Fig. 3 shows a conventional split-architecture audio codec interfacing to a device such as a low speed voice band modem **510** in accordance with the AC '97 Specification.

In particular, in Fig. 3, an AC controller sub-system **700**
20 interfaces to an AC analog sub-system **702** via a five-wire synchronized serial data bus (i.e., a time division multiplexed (TDM) bus) referred to as the AC link **504**. The five-wire TDM bus of the AC link **504** comprises a sync signal **712**, a reset signal **520**, a serial TDM data stream SDATA_OUT **716** from the AC controller sub-system **700** to the AC
25 analog sub-system **702**, a clock signal BIT_CLK **714**, and a serial TDM data stream SDATA_IN **718** from the AC analog sub-system **702** to the AC controller **700**. The clock signal BIT_CLK **714** is derived by a clock source **506** in or relating to the controller **700**.

The circuitry in the conventional AC analog sub-system **702**
30 which interfaces to an external analog device such as the low-speed voice

band modem includes an analog-to-digital converter (ADC) **522** and a digital-to-analog converter (DAC) **524**. The ADC **522** samples the analog modem signal input to the AC analog sub-system **702** and provides 16-, 18- or 20-bit data at 48 Ks/s to the AC link **504** for insertion into an appropriate time slot (e.g., time slot 5) of the serial TDM data stream **SDATA_IN 718** input to the AC controller sub-system **700**. Conversely, the DAC **524** receives 16-, 18- or 20-bit data from the serial TDM data stream **SDATA_OUT 716** from the AC controller sub-system **700** of the AC link **504** and converts the same into an analog signal output to the low-speed voice band modem **510**. Conventional demodulation and modulation techniques such as quadrature amplitude modulation (QAM) or Carrierless Amplitude and Phase (CAP) may be performed by a digital signal processor (DSP) and/or other processor in conjunction with the ADC **522** and/or DAC **524**.

Fig. 4 depicts a conventional sync signal **712**, serial TDM data stream **SDATA_OUT 716**, and serial TDM data stream **SDATA_IN 718**, in a twelve slot TDM bi-directional data stream between the analog and controller sub-systems **702**, **700** of a split-architecture audio codec such as in accordance with the AC '97 Specification. The twelve time slots 1 to 12 of the serial TDM data streams **SDATA_OUT 716** and **SDATA_IN 718** are framed by the conventional sync signal **712**. The sync signal **712** is derived from a TAG Phase **600** received during time slot 0. All time slots are 20 bits wide.

The sync signal **712** synchronizes the reception and transmission of the **SDATA_OUT 716** and **SDATA_IN 718** with respect to clock signal **BIT_CLK 714**. This synchronization between the data lines and the clock signal is shown in more detail in Fig. 5.

In particular, Fig. 5 shows the clock signal **BIT_CLK 714** and serial TDM data stream **SDATA_OUT 716** with reference to the sync

signal **712**. The sync signal **712** is based on the clock signal BIT_CLK **714**, which is a fixed 12.288 MHz clock signal.

Fig. 6 shows an implementation of the AC analog sub-system (i.e., codec) **702** interfaced with the controller **700** on a low voltage circuit side using a conventional differential implementation of the serial interface. As shown, the codec **702** is typically exposed to voltages in excess of the power voltage, and therefore is referred to herein as a high voltage circuit. In some situations it is desirable to AC couple a clock signal in a serial interface such that a codec or other high voltage circuit **702** may be electrically isolated from the ground of a low voltage circuit **700**. It would similarly be desirable to AC couple the transmit data signal **716**, the receive data signal **718**, the frame sync signal **712**, and the reset signal **520**. If all signals between the low voltage controller **700** and the high voltage codec **702** are AC coupled, then there is essentially no need for a connection to exist between the ground of the low voltage controller **700** and the ground of the high voltage codec **702**.

Unfortunately, in practical situations, once the grounds between the low voltage controller **700** and high voltage codec **702** are broken, a large common mode voltage may exist between the ground potential of the low voltage controller **700** and the ground potential of the high voltage codec **702**. This large common mode voltage may interfere with the AC coupled digital signals in the isolated high voltage codec **702**. Moreover, the cost of an isolating transformer **791** can be significant, and if the codec **702** is to perform impedance emulation with a central office, the transformer **791** can degrade impedance matching between the codec **702** and the telephone line.

Thus, an alternative implementation of the serial interface has been developed wherein the codec (which may be connected to a telephone line, modem, audio source, etc.) is placed in the high voltage section of the system as shown in Fig. 7 to eliminate the expensive and

large transformer 791 which is otherwise traditionally used to couple the low voltage side to the high voltage side. This technique eliminates the need for the transformer 791, but is disadvantageous for other reasons, e.g., because it requires additional hardware such as the AC coupling capacitors C (typically rated at 3000 V AC).

An example of such a serial interface could be used with a LUCENT TECHNOLOGIES CSP1034 multi-processor mode SIO interface. In such an example, five serial lines are typically needed to provide the interface, and each of the five signals corresponding to the five serial lines would need to be converted to differential signal pairs. Each of the five signal pairs would each need to be isolated with a corresponding pair of capacitors for voltage isolation, requiring a total of ten (10) high voltage capacitors.

It is important to reduce the number of communication lines necessary to interface between circuits, particularly where one of the circuits is subject to higher voltages, e.g., a codec, because of the relative cost of the hardware and space required for each of the individual lines. Thus, there is a need for reducing the number of signal lines in a synchronous serial interface such as the time division multiplex (TDM) serial interface of the AC '97 specification.

SUMMARY OF THE INVENTION

In accordance with the principles of the present invention, a synchronizing data protocol for use in occasionally synchronizing timing between a master device and a slave device comprises a preamble insertion module in the master timing device adapted to insert a preamble code word into a data stream for transmission to the slave device. A synchronizing preamble detection module in the slave device is adapted to detect a presence of the preamble code word in the data stream.

A method of synchronizing a slave device to a master device over a serial data bus in accordance with another aspect of the present invention comprises providing an interrupt signal to the slave device. The data stream received by the slave device is monitored for a presence of a
5 synchronizing preamble code word. A timing in the slave device is based on a timing of a detection of the synchronizing preamble code word by the slave device.

BRIEF DESCRIPTION OF THE DRAWINGS

10 Features and advantages of the present invention will become apparent to those skilled in the art from the following description with reference to the drawings, in which:

Fig. 1 shows an exemplary synchronizing data protocol used without the need for frame synchronization, in accordance with the
15 principles of the present invention.

Fig. 2 is a more detailed diagram of the synchronizing data protocol shown in Fig. 1.

Fig. 3 shows a conventional four signal (plus reset) serial interface between two separate circuits in accordance with the AC '97
20 Specification.

Fig. 4 depicts a conventional sync signal, and twelve slot, serial TDM transmit and receive data streams between the analog and controller sub-systems of a split-architecture audio codec such as is shown in Fig. 3.

25 Fig. 5 shows in greater detail an implementation of a bit clock signal and serial TDM data stream with reference to the sync signal used with the serial interface shown in Fig. 3.

Fig. 6 shows a conventional transformer isolation circuit.

Fig. 7 shows a conventional digital AC coupling capacitor
30 isolation circuit.

DETAILED DESCRIPTION OF ILLUSTRATIVE EMBODIMENTS

The present invention provides a synchronizing data protocol comprising one or more serial input-output (SIO) control word(s) and data passed across a high voltage interface, which eliminates the need for a frame synchronization signal.

The traditional method of passing codec data across a high voltage interface was shown and described with respect to Fig. 6, i.e., by placing a codec **702** on the low voltage side of an interface and passing analog signals across the high voltage barrier with high voltage isolation (e.g., 3000 volt isolation) provided by a transformer **791**. The transformer **791** would also isolate the codec **702** from low frequency common mode noise, e.g., riding on the twisted pair wiring of a telephone line **104**. In addition, the transformer **791** would isolate the telephone system of a central office **113** from high frequency common mode noise generated by the digital subsystems in the controller **700** and codec **702**. However, the use of a transformer **791** makes impedance matching by the codec **702** with the central office **113** more difficult due to amplitude and phase distortion introduced by the transformer **791**.

Alternatively, to eliminate the need for the transformer **791**, the codec **702** may be placed on the high voltage circuit side as shown in Fig. 7. In this case, the codec **702** is AC coupled to the differential digital SIO signals **712**, **714**, **716**, **718** and **520**. Using this technique, the differential receivers pass digital signals and reject the majority of common mode noise, and the combination of high voltage capacitors **C** and differential receivers replace the functions otherwise provided by the transformer **791** (Fig. 6).

The principles of the present invention further the conventional technique shown in Fig. 7 by eliminating the need for the sync signal **712**. Elimination of the sync signal **712** saves two pins on

each of the controller **700** and the codec **702**, eliminates required board space, and reduces overall system cost, e.g., by eliminating two expensive AC coupling capacitors **C**.

In accordance with the disclosed embodiment of the present invention shown in Fig. 1, elimination of the conventional sync signals **712** is accomplished by adding circuitry or software on one side of the interface, e.g., on the low voltage side, to insert a predetermined synchronizing code word into a data stream and, on the high voltage side, to recognize the presence of the predetermined synchronizing code word.

While the described embodiment relates to a synchronization controlled by the device on the low voltage side, with the device on the high voltage side synchronizing to the clocking of the device on the low voltage side, the principles of the present invention relate to synchronization controlled by either the device on the low voltage side (e.g., the controller **100**) or the device on the high voltage side (e.g., the codec **106**).

Moreover, while the disclosed embodiments relate to a controller and audio codec communicating over an AC link in general conformance with the AC '97 Specification, the principles of the present invention relate equally to any serial communication link otherwise requiring a synchronization pulse such as a framing pulse, e.g., in a synchronous serial data stream or in a time division multiplexed (TDM) synchronous serial data stream communication system.

In accordance with the disclosed embodiment, a data protocol is implemented which replaces the need for a frame synchronization signal. The data protocol is particularly useful with very stable TDM communication lines.

In the disclosed embodiment, 1-bit sigma delta data is transferred between a controller and an audio codec. A data protocol is implemented over the data line which provides occasional synchronizing between the data being transmitted by a transmitting system (e.g., the

codec **106** in Fig. 1) and the data being received by a receiving system (e.g., the controller **100** in Fig. 1).

The principles of the present invention are further applicable to implementations which further reduce the required signals between digital components, e.g., between the controller **100** and the codec **106**. For instance, the required number of signals and corresponding AC coupling capacitors may be further reduced by combining a synchronizing data protocol in accordance with the principles of the present invention with a signal encoding the transmit signals and data clock together using bi-phase encoding, as described in U.S. Appl. No. 09/013,943 entitled "Method and Apparatus For Combining Serial Data With A Clock Signal" filed January 27, 1998, the specification of which is expressly incorporated herein by reference in its entirety.

Fig. 1 shows an embodiment of a serial TDM interface forming the basis of an AC Link in accordance with the AC '97 Specification. However, in Fig. 1, the otherwise conventional frame synchronization signal (e.g., **712** in Fig. 7) is eliminated and replaced by a synchronizing data protocol in the controller **100** in accordance with the principles of the present invention.

In particular, Fig. 1 shows implementation of a codec **106** on a high voltage circuit side, and a controller **100** on a low voltage circuit side. In the disclosed embodiment, the codec **106** is interfaced with a DAA **102**, which in turn is connected through a telephone line **104** to a telephone company central office **113**. However, in accordance with the principles of the present invention and the AC '97 Specification, the codec **106** may be interfaced with virtually any analog device, particularly those related to consumer audio and telephonic devices.

The serial interface between the controller **100** and the codec **106** comprises three basic communication signals, i.e., a data transmit TDM signal **116**, a data receive TDM signal **118**, and a data clock

signal **110**. If desired, a reset signal (not shown) may also be implemented based on the needs of the particular application.

In the disclosed embodiment, the transmit signal **116**, the receive signal **118**, and the clock signal **110** are each differential signals including at least one AC coupling capacitor **C** in each side of the respective differential signal line. However, the principles of the present invention are equally applicable to single-ended serial lines. Moreover, the transmit signal **116** and the clock signal **110** may be combined into a single encoded signal as described in U.S. Appl. No. 09/013,943, the specification of which is explicitly incorporated herein by reference.

The low voltage circuit, i.e., the controller **100** includes a preamble insertion module **134**, a clock absence timer **132**, and a clock source **130**. The high voltage circuit, i.e., the codec **106**, includes a sync preamble detect module **140**, a clock absence timer **138**, a synchronized timing module **136**, and a buffer **142** which is activated upon detection of the preamble code word in the data stream following an absence of the clock signal.

In particular, the preamble insertion module **134** is adapted to insert a predetermined, preamble code word, e.g., '11101', into a data stream transmitted on the transmit signal line **116**. Depending upon the particular application, the preamble code word may be any code and may be of any length. Preferably, the preamble code word is at least five bits long, and preferably starts with a '111'.

The preamble code word is inserted into the data stream following an absence of the clock signal. Thus, to initiate a re-synchronization, the controller **100** intentionally halts the clock source **130** for a predetermined period of time, e.g., for at least 700 mS, then inserts the preamble code word into the transmit data stream **116**.

The clock absence timer **132** in the controller **100** relates to a self-monitoring function by the controller **100** to determine how long the

controller **100** has halted a clock source **130** from being output on the clock signal line **110**. After a sufficient amount of time that the clock is withheld from the codec **106**, e.g., after 700 mS, the preamble insertion module **134** inserts the preamble code word into an appropriate time slot
5 of the data stream being transmitted on the transmit signal line **116**.

The controller **100** is a master and the codec **106** is a slave with respect to the timing of the serial TDM interface. Thus, in the codec, timing is received and synchronized by the synchronized timing module **136**. The clock signal is also monitored by the clock absence timer **138**.
10 The clock absence timer **138** activates the sync preamble detect function **140** upon detection of the absence of the clock signal for more than a threshold amount of time, e.g., for more than 700 mS. Upon activation, the sync preamble detect function **140** monitors the data stream received on the transmit line **116** for the presence of the predetermined preamble
15 code word.

The present inventors realized the stability of TDM serial interfaces in general, and the AC link in particular, in development of the present invention. For instance, it was empirically determined that a typical bit error rate (BER) of data passing over a conventional transmit
20 signal line **716** (Fig. 7) was on the order of less than 1 bit error every ten (10) years. Thus, it was determined, with such stability in the serial TDM data communications that frame synchronization need not occur every frame to maintain such stability. It was determined that an occasional synchronization carried out through the recognition of a unique data
25 protocol would provide adequate synchronization to maintain an extremely stable communication link (i.e., having a low bit rate), while at the same time would eliminate the cost and space required for the additional synchronization signal (e.g., **712** in the conventional interface shown in Fig. 7). In accordance with the principles of the present invention, after
30 synchronization, the data communications would be allowed to free run

using only the transmit signal **116**, the receive signal **118**, and the clock signal **110**.

Fig. 2 shows an embodiment of the sync preamble detect function **140** wherein control information may be passed from the controller **100** to the codec **106** together with the preamble code word.

In particular, a data stream buffer **210** may be parsed apart to determine the presence of the preamble code word together with address and/or data information. The preamble code word **202** may be, e.g., an 8-bit code word such as '11000110', and may be received in a first buffer **210c**. A second buffer **210b** may receive address information, e.g., up to 256 registers each containing an 8-bit address. Similarly, general data information, e.g., 16-bit data words, may be received in a third buffer **210a**.

Upon recognition of the preamble code word **202**, a comparator and latch controller **204** may clock the data contained in the data buffer **210a** into an appropriate register **208**, and the address(es) contained in the address buffer **210b** into appropriate registers(s) **206**. The data and address information latched into the registers **206**, **208** may relate to system parameters typically initiated at system startup or at timing synchronization.

To distinguish control words from signal data, an interrupt procedure may be set up to eliminate the need for an interrupt line. In the disclosed embodiment, this interrupt procedure is triggered by the absence of the clock signal as detected by the clock absence timer **138**. When the clock stops for at least a predetermined length of time, the sync preamble detect function **140** is reset to thus monitor for the presence of the preamble code word **202**. Once the preamble code word **202** is detected, the data and address will be latched into the appropriate registers **206**, **208**. The data and address registers **206**, **208** will not

accept new data until another interrupt is received, i.e., until the clock stops and starts again.

Once all of the desired registers are all programmed, a bit **143** is set to enable the buffer **142** enabling data, e.g., sigma delta data transfer to the codec. As long as the clock signal continues to be present at the codec **106**, data is transferred to the codec filters.

Thus, in accordance with the principles of the present invention, the synchronizing data protocol need only be transmitted once in certain applications, at the beginning of a communication transaction. However, as dictated by the particular application, it may be desirable to occasionally activate the synchronizing data protocol to further ensure reliable and error free communications. For instance, upon detection of an absence of a clock signal received by the codec **106** of at least a minimum amount of time, e.g., an absence of a clock signal for greater than 25 microseconds (uS), a warm synchronization may be activated wherein it is likely that stored registers and other memory resident parameters remain stable. In such a warm synchronization, the preamble is inserted into the data stream by the preamble insertion module **134** in the controller **100**, the preamble is detected by the sync preamble detector **140** in the codec **106**, and data communications proceed therefrom.

Moreover, in another embodiment, a loss of powering current from a telephone line may be sensed. In certain countries, current available on a telephone line may routinely be interrupted for up to 700 milliseconds (mS), thus circuitry being powered by line current must accommodate for such gaps in powering current. However, in the event of an out-of-specification current interruption of greater than 700 mS, the codec **106** and other circuitry (if being powered by the telephone line **104**) may experience an unpredictable period without power. In such a case, upon recovery from the current loss, a synchronizing data protocol in

accordance with the principles of the present invention may be implemented to re-synchronize communications.

In testing, replacement of a synchronizing signal with a synchronizing data protocol in accordance with the principles of the present invention has proved very reliable, even with communication lines which exhibit a large amount of noise.

The principles of the present invention have applicability with synchronous serial data streams in general, and not just with time division multiplexed serial data streams. For instance, the invention may be implemented with a continuous synchronous serial transmission data stream otherwise conventionally utilizing a framing signal or other signal to synchronize a receiving device with the serial data transmission.

Moreover, while the present invention is described with reference to a specific codec, i.e., an AC '97 conforming codec, the principles of the present invention relate to communication with any codec device.

While the invention has been described with reference to the exemplary preferred embodiments thereof, those skilled in the art will be able to make various modifications to the described embodiments of the invention without departing from the true spirit and scope of the invention.